

A CASE STUDY OF OPTIMAL INPUT-OUTPUT SYSTEM WITH SAMPLED-DATA CONTROL: DING ET AL. FORCE AND FATIGUE MUSCULAR CONTROL MODEL

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ABSTRACT. The objective of this article is to make the analysis of the muscular force response to optimize electrical pulses train using Ding et al. force-fatigue model. A geometric analysis of the dynamics is provided and very preliminary results are presented in the frame of optimal control using a simplified input-output model. In parallel, to take into account the physical constraints of the problem, partial state observation and input restrictions, an optimized pulses train is computed with a model predictive control, where a non-linear observer is used to estimate the state-variables.

1. Introduction. Functional Electrical Stimulation (FES) consists of applying an electrical stimulation to the muscle, in order to produce functional movements. It can be used for the muscular reinforcement, reeducation of the muscle and in the case of paralysis to activate the paralyzed muscles. Mathematically FES leads to a sampled-data control problem which can be analyzed in this framework.

The simulations of muscular response to electrical stimulations are based on dynamics models. The origin comes from the Hill-Langmuir equation in the context of biochemistry and pharmacology, see [12]. More recent models in the framework of model identification in non-linear control are due to Bobet and Stein [2], Law and Shields [15] and a more sophisticated model was proposed by Ding et al. [6, 7, 8, 9] where the force model is coupled to a fatigue model based on experimental results [20]. This led to a set of five differential equations with a sampled-data control which can be used to describe the force response to a pulses train of electrical stimulations and we shall refer to this model as the Ding et al. force-fatigue model in the sequel.

There is a limited literature that used this model to design an optimized train of pulses to control the force level [6, 7, 8, 9, 1]. Our aim is to produce a more

2010 *Mathematics Subject Classification.* 49K15, 93B07, 92B05.

Key words and phrases. Muscular force and fatigue model, optimization of sampled-data control, non-linear observer, model predictive control.

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complete study of the problem in the framework of non-linear optimal control, using sampled-data controls.

A geometric analysis of the model is provided and preliminary results are presented in the framework of optimal control with sampled-data control [3]. It is based on a simplified dynamics using the geometric properties of the force-fatigue model and control reduction to simplify the physical control constraints. The complete system is analyzed in details using a control predictive strategy (MPC), see [17, 19] coupled with a non-linear observer based on [11] to estimate the state variables.

The article is organized as follows. In Section 2, we make a brief presentation of Ding et al. force-fatigue model based on [20]. In Section 3, the dynamics of the force model is investigated to describe the input-output properties. In Section 4, the force-fatigue model is analyzed in the framework of geometric optimal sampled-data control systems and preliminary results are presented with a simplified model using a model reduction and an input transformation. Section 5 is devoted to the observer description. In Section 6, MPC method is presented using a further discretization of the dynamics to conclude with the algorithm implemented to compute in practice the optimized pulses trains. Numerical results are presented in the final Section 7.

2. Mathematical force-fatigue model. We refer to [20] for a complete description and discussions of the model. The Ding et al. model studied in this article is presented next in the framework of model dynamics construction based on the so-called *tetania* phenomenon in muscular responses. The first part of the model is the force response (output) to electrical stimulations pulses (input). The pulses are normalized Dirac impulses at times $0 = t_1 < t_2 < \dots < t_n$:

$$v(t) = \sum_{i=1}^n \delta(t - t_i)$$

and $I_i = t_i - t_{i-1}$ is the interpulse and convexifying leads to apply the input:

$$v(t) = \sum_{i=1}^n \eta_i \delta(t - t_i).$$

with parameters $\eta_i \in [0, 1]$, $i \in \{1, \dots, n\}$. Such pulses train feeds a first-order model to produce an output according to the dynamics

$$\dot{u}(t) + \frac{u(t)}{\tau_c} = \frac{1}{\tau_c} \sum_{i=1}^n R_i \eta_i \delta(t - t_i) \quad (1)$$

where R_i is a scaling factor associated to the phenomenon of tetania and corresponds to an accumulated effect of successive pulses and is modelled as

$$R_i = \begin{cases} 1 & \text{for } i = 1 \\ 1 + (R_0 - 1) \exp\left(-\frac{t_i - t_{i-1}}{\tau_c}\right) & \text{for } i > 1 \end{cases}$$

where the magnitude is characterized by R_0 which is the limit case to high-frequency pulse. The FES response to the input is denoted by $E_s(t)$ and is given by

$$E_s(t) = \frac{1}{\tau_c} \sum_{i=1}^n R_i \eta_i H(t - t_i) \exp\left(-\frac{t - t_i}{\tau_c}\right) \quad (2)$$

where $H(t - t_i) = \begin{cases} 0 & \text{if } t < t_i \\ 1 & \text{if } t \geq t_i \end{cases}$ is the Heaviside function.

The force response to such a train is modelled by the two equations of the so-called *force model*:

$$\frac{dC_N}{dt}(t) + \frac{C_N(t)}{\tau_c} = E_s(t) \quad (3)$$

which corresponds to a first order (resonant) linear dynamics which can be integrated with $C_N(0) = 0$ and the force response $F(t)$ is described by the equation

$$\frac{dF}{dt}(t) = A(t) a(t) - F(t) b(t) \quad (4)$$

where A is a fatigue variable. Non-linear features of the model are described by the two mappings a, b :

$$a(t) = \frac{C_N(t)}{K_m(t) + C_N(t)}, \quad b(t) = \frac{1}{\tau_1(t) + \tau_2 a(t)} \quad (5)$$

where K_m, τ_1 are fatigue variables and τ_2 an additional parameter.

The complete model is obtained by describing the evolutions of the variables associated to *fatigue* and corresponds to the *linear dynamics*

$$\frac{dA}{dt}(t) = -\frac{A(t) - A_{rest}}{\tau_{fat}} + \alpha_A F(t) \quad (6)$$

$$\frac{dK_m}{dt}(t) = -\frac{K_m(t) - K_{m,rest}}{\tau_{fat}} + \alpha_{K_m} F(t) \quad (7)$$

$$\frac{d\tau_1}{dt}(t) = -\frac{\tau_1(t) - \tau_{1,rest}}{\tau_{fat}} + \alpha_{\tau_1} F(t). \quad (8)$$

The full set of equations (3)-(4)-(6)-(7)-(8) are the *force and fatigue model*.

We refer to Table 1 for the definitions and details of the symbols of the force-fatigue model.

For the purpose of our analysis the force-fatigue model is written as the single-input control system:

$$\frac{dx}{dt}(t) = F_0(x(t)) + u(t) F_1(x(t)) \quad (9)$$

with $x = (x_1, x_2, x_3, x_4, x_5) = (C_N, F, A, K_m, \tau_1)$ where u denotes the control $u = E_s(t)$ corresponding to the *sampled physical control data*:

$$(I_1, I_2, \dots, I_n, \eta_1, \dots, \eta_n) \quad (10)$$

with constraints

$$I_{\min} \leq I_i \leq I_{\max}, \quad 0 \leq \eta_i \leq 1$$

corresponding to interimpulses bounds and amplitude convexification. This leads to a *non-linear model* with sampled control data with prescribed *convex control constraints* $(\eta, I) \in C$; $\eta = (\eta_1, \dots, \eta_n)$, $I = (I_1, \dots, I_n)$.

3. The force model. The force model can be briefly investigated to describe preliminary results.

TABLE 1. Margin settings

Symbol	Unit	Value	description
C_N	—	—	Normalized amount of Ca^{2+} -troponin complex
F	N	—	Force generated by muscle
t_i	ms	—	Time of the i^{th} pulse
n	—	—	Total number of the pulses before time t
i	—	—	Stimulation pulse index
τ_c	ms	20	Time constant that commands the rise and the decay of C_N
R_0	—	1.143	Term of the enhancement in C_N from successive stimuli
A	$\frac{N}{ms}$	—	Scaling factor for the force and the shortening velocity of muscle
τ_1	ms	—	Force decline time constant when strongly bound cross-bridges absent
τ_2	ms	124.4	Force decline time constant due to friction between actin and myosin
K_m	—	—	Sensitivity of strongly bound cross-bridges to C_N
A_{rest}	$\frac{N}{ms}$	3.009	Value of the variable A when muscle is not fatigued
$K_{m,rest}$	—	0.103	Value of the variable K_m when muscle is not fatigued
$\tau_{1,rest}$	ms	50.95	The value of the variable τ_1 when muscle is not fatigued
α_A	$\frac{1}{ms^2}$	$-4.0 \cdot 10^{-7}$	Coefficient for the force-model variable A in the fatigue model
α_{K_m}	$\frac{1}{msN}$	$1.9 \cdot 10^{-8}$	Coefficient for the force-model variable K_m in the fatigue model
α_{τ_1}	$\frac{1}{N}$	$2.1 \cdot 10^{-5}$	Coefficient for force-model variable τ_1 in the fatigue model
τ_{fat}	s	127	Time constant controlling the recovery of (A, K_m, τ_1)

3.1. Parameterization. Observe that in the force model the static non-linearities described by the mappings a and b can be absorbed by time reparameterization to provide an explicit form for the force responses.

Proposition 1. *The force model can be integrated by quadratures using a time reparameterization.*

Proof. Integrating the linear system (3), the equation (4) can be written as

$$\frac{dF}{ds}(s) = c(s) - F(s) \quad (11)$$

with $ds = b(t) dt$, $c(s) = A(s)a(s)/b(s)$. It can be integrated using Lagrange formula. \square

Thus this gives an explicit force response in the time parameter s

$$(I, \eta, s) \rightarrow F(I, \eta, s). \quad (12)$$

Clearly we have

Lemma 3.1. *The above mapping is smooth with respect to I, η and piecewise smooth with respect to s .*

3.2. Smoothing process. For the sake of providing a *smooth response* for the observer it is sufficient to smooth the physical sampled control data as follows: use a bump function to smooth each Heaviside mapping $H(t - t_i)$ at the sampling time t_i .

3.3. Input-simplification. For the sake of the geometric analysis of the dynamics and to simplify the control constraints on (I, η) the FES signal $E_s(t)$ is taken as the input $u(\cdot)$ of the control system. Using (2) one can write

$$u(t) = E_s(t) = e^{-t/\tau_c} v(t)$$

with

$$v(t) = \frac{1}{\tau_c} \sum_{i=1}^n H(t - t_i) R_i \eta_i e^{t_i/\tau_c}. \quad (13)$$

Whence $0 = t_1 < t_2 < \dots < t_n$ are given $v(t)$ is a piecewise constant control depending upon the parameters η_1, \dots, η_n and the dynamics of the force model can be analyzed in the frame of geometric control.

4. Force-fatigue control model. First of all the control system (9) is analyzed in the framework of geometric control.

4.1. The concepts of geometric control system [13, 14]. Consider a (smooth) control single-input control system.

$$\begin{cases} \frac{dx}{dt} = X(x) + uY(x) \\ y = h(x) \end{cases} \quad (14)$$

where $x \in \mathbb{R}^n$, $y = h(x) \in \mathbb{R}$ corresponds to a (smooth) single-observation mapping. The following concepts rely on seminal results of geometric control, see [18, 13].

We denote $[U, V]$ the Lie bracket of two (smooth) vector fields of \mathbb{R}^n :

$$[U, V](x) = \frac{\partial U}{\partial x}(x) V(x) - \frac{\partial V}{\partial x}(x) U(x)$$

and a vector field U acts on (smooth) mappings f with the Lie derivative:

$$\mathcal{L}_U f = \frac{\partial f}{\partial x} U(x).$$

Let $D^1 = \text{span}\{X, Y\}$ and define recursively : $D^k = \text{span}\{D^k \cup [D, D^{k-1}]\}$, $k > 1$. The length of $[X_{i_1}, \dots, [X_{i_{k-1}}, X_{i_k}] \dots]$ is k . Hence D^k represents Lie brackets of X, Y with lengths smaller than $k + 1$ and denote $D_{L.A.} = \text{span} \cup_{k \geq 0} D^k$ the *Lie*

algebra generated by $\{X, Y\}$. The system is called *weakly controllable* if for each $x \in \mathbb{R}^n$, $D_{L.A.}(x) = \mathbb{R}^n$.

The *observation space* is the set of mappings: $\Theta = \{\mathcal{L}_G h; G \in D_{L.A.}\}$ and the system is called *observable* if for each $x_1, x_2 \in \mathbb{R}^n$, $x_1 \neq x_2$ there exists $\theta \in \Theta$ such that $\theta(x_1) \neq \theta(x_2)$.

Taking $x_0 \in \mathbb{R}^n$, a *frame* at x_0 is a set of elements X_1, \dots, X_n of $D_{L.A.}$ such that: X_1, \dots, X_n are linearly independent at x_0 and $\sum_{i=1}^n \text{length}(X_i)$ is minimal.

The system is called *feedback linearizable* in an open set $\mathcal{V} \subset \mathbb{R}^n$ if $\dot{x} = X(x) + uY(x)$ is feedback equivalent to the linear system $\dot{x} = Ax + ub$, that is there exists a diffeomorphism φ on \mathcal{V} , $\varphi(0) = 0$ and a feedback $u = \alpha(x) + \beta(x)v$, $\beta(x) \neq 0$ such that $g \cdot (X, Y) = (A, b)$ with $g = (\varphi, \alpha, \beta)$ acting by change of coordinates and (affine) feedback.

Fix $x(0) = x_0$ and $T > 0$ and consider the *extremity mapping*: $E^{x_0, T} : u(\cdot) \in L^\infty([0, T]) \mapsto x(T, x_0, u)$ where $x(\cdot)$ is the response of $\dot{x} = X(x) + uY(x)$ to u defined on $[0, T]$.

The control $u(\cdot)$ is called *singular* on $[0, T]$ if the extremity mapping $E^{x_0, T}$ is not of maximal rank n when evaluated at $u(\cdot)$.

Geometric analysis of an observed system of the form (14) amounts to compute $D_{L.A.}$, the observation space, the singular controls and the feedback equivalence properties. Achievements of geometric optimal control amounts to synthesize the optimal control in relation with the Lie brackets properties of a frame.

4.2. Optimal control a force-fatigue model with sampled control data.

4.2.1. *Concepts.* The control in the force-fatigue model (1) falls into the framework of sampled-optimal control problem and we refer to [3] for more details. We use the following terminology.

Consider a control system $\dot{x} = f(x, u)$. When the state $x(\cdot)$ and the control $u(\cdot)$ evolve continuously in time, we speak of a continuous-time control problem and the control is said *permanent*. The *sampled-data control* case is when $u(\cdot)$ is piecewise constant.

4.2.2. *Application to the force-fatigue model.* For simplicity consider the reduced model in dimension 3 related to the force-fatigue model.

Let the minimization problem be defined by

$$\int_0^T (u(s)^2 + (F - F_{ref})^2) ds \rightarrow \min_{|u| \leq 1}$$

subject to the dynamics in $\tilde{x} = (C_N, F, A)$ given by

$$\begin{aligned} \frac{dC_N}{dt} &= -\frac{C_N}{\tau_c} + u \\ \frac{dF}{dt} &= aA - bF \\ \frac{dA}{dt} &= -\frac{A - A_{rest}}{\tau_{fat}} + \alpha_A F \end{aligned} \tag{15}$$

and with the initial conditions

$$F(0) = A(0) = C_N(0) = 0.$$

The parameters $T, \tau_2, \tau_{fat}, \alpha_A, A_{rest}$ and the variables K_m, τ_1 are constant and fixed to some prescribed values.

The system (15) can be written into the form

$$\dot{\tilde{x}} = \tilde{F}_0(\tilde{x}) + u \tilde{F}_1(\tilde{x}).$$

Remark 1. The model (15) is a simplification of the force-fatigue model. In particular, without loss of generality, we don't take into account the factor $\exp(-t/\tau_c)/\tau_c$ appearing in (2). Control constraints $|u| \leq 1$ are not the physical constraints, see Section 3.3. Also the fatigue dynamics is reduced to a single equation, motivated by the controllability properties of the system (6)-(7)-(8).

The permanent control case. The problem can be summarized as a permanent optimal control problem as follows

$$\begin{cases} \min_{u(\cdot)} \int_0^T (u(s)^2 + (F - F_{ref})^2) ds \\ \dot{\tilde{x}} = \tilde{F}_0(\tilde{x}) + u \tilde{F}_1(\tilde{x}) \\ u(t) \in [-1, 1] \\ (C_N(0), F(0), A(0)) = (0, 0, 0). \end{cases} \quad (16)$$

The pseudo-Hamiltonian of the system is

$$H(\tilde{x}, p, p^0, u) = p \cdot (\tilde{F}_0(\tilde{x}) + u \tilde{F}_1(\tilde{x})) + p^0 (u^2 + (F - F_{ref})^2)$$

where $(p, p^0) : [0, T] \mapsto \mathbb{R}^4$ is the adjoint vector.

We denote by $H_i = p \cdot \tilde{F}_i$, $i = 0, 1$ the Hamiltonian lifts of the vector fields \tilde{F}_i , $i = 0, 1$.

Normal case: $p^0 = -1/2$. Applying the Pontryagin maximum principle (PMP), the optimal (permanent) control is given in the normal case by

$$u(t) = \begin{cases} 1 & \text{if } H_1(\tilde{x}(t), p(t)) \geq 1, \\ -1 & \text{if } H_1(\tilde{x}(t), p(t)) \leq -1, \\ H_1(\tilde{x}(t), p(t)) & \text{otherwise.} \end{cases} \quad (17)$$

Abnormal case: $p^0 = 0$. Abnormal controls are characterized by the equation $p \cdot \tilde{F}_1 = 0$. Differentiating twice with respect to t leads to

$$p \cdot [\tilde{F}_0, \tilde{F}_1] = 0, \quad p \cdot \left([[\tilde{F}_0, \tilde{F}_1], \tilde{F}_0] + u [[\tilde{F}_0, \tilde{F}_1], \tilde{F}_1] \right) = 0.$$

Eliminating p we obtain $D + u D' = 0$ where

$$\begin{aligned} D &= \det(\tilde{F}_1, [\tilde{F}_1, \tilde{F}_0], [[\tilde{F}_1, \tilde{F}_0], \tilde{F}_0]) \\ D' &= \det(\tilde{F}_1, [\tilde{F}_1, \tilde{F}_0], [[\tilde{F}_1, \tilde{F}_0], \tilde{F}_1]). \end{aligned}$$

Computing we have

$$D = -\alpha_A (a'(C_N) A - b'(C_N) F)^2, \quad D' = 0$$

and $D = 0$ is equivalent to $A(C_N(\tau_1 + \tau_2) + K_m \tau_1)^2 + F \tau_2 (C_N + K_m)^2 = 0$. This implies $F = A = 0$ and the following lemma.

Lemma 4.1. *There are no physically admissible singular trajectories for the problem (15).*

The sampled-data control case. The corresponding optimal sampled-data control problem is given by

$$\begin{cases} \min \int_0^T (u(kT_s)^2 + (F - F_{ref})^2) dt, & \text{with } k = \lfloor t/T_s \rfloor \\ \dot{\tilde{x}} = \tilde{F}_0(\tilde{x}) + u \tilde{F}_1(\tilde{x}) \\ u(kT_s) \in [-1, 1] \\ (C_N(0), F(0), A(0)) = (0, 0, 0). \end{cases} \quad (18)$$

where $T_s > 0$ is a *fixed* sampling period such that $T = jT_s$ for some $j \in \mathbb{N}$.

Normal case: $p^0 = -1/2$. Following [3], the optimal (sampled-data) control is

$$u(kT_s) \in \arg \max_{y \in [-1, 1]} \frac{1}{T} \int_{kT_s}^{(k+1)T_s} H(s, \tilde{x}(s), p(s), p^0, y) ds \quad (19)$$

for all $k \in \{0, \dots, d\}$.

Write

$$\bar{H}_1 = \frac{1}{T} \int_{kT_s}^{(k+1)T_s} p_1(s) ds$$

then, the optimal sampled-data control is

$$u(kT_s) = \begin{cases} 1 & \text{if } \bar{H}_1 > 1, \\ -1 & \text{if } \bar{H}_1 < -1, \\ \bar{H}_1 & \text{otherwise.} \end{cases} \quad (20)$$

Numerical results. In Fig.1, we represent numerical results for several values of T/T_s . The optimal permanent control is represented with thin continuous line. A Gaussian quadrature rule is used to compute \bar{H}_1 . We observe the convergence of the optimal sampled-data control to the permanent control as T_s tends to 0.

5. Observer.

5.1. Sensibility study of the force versus K_m . The force evolution is compared for $K_{m,rest}$ and different values $K'_{m,rest}$ ($\pm 30\%$ of error) for $I = 10ms$, $50ms$ and $100ms$ (see Fig. 2 for $I = 10ms$). In the case of $I = 10ms$, the maximum force error (see Fig. 3) is of -0.3% . Following interpulse value, the maximum force error is obtained for $I = 100ms$ (-1.3%) which means that a tolerance of $\pm 30\%$ gives force evolution with a good accuracy.

5.2. High-gain observer synthesis for the estimation of A and τ_1 . In this section, we design a modified version of the standard high-gain observer given in [11] taking into account the specific structure of the problem. The system defined by the force equation (4) and the fatigue model (6) and (8) can be rewritten as the single input-output system:

$$\begin{cases} \dot{x}(t) = a^m(t, E_s(t))f(x(t), E_s(t)) \\ y(t) = h(x(t)) = F(t), \end{cases} \quad (21)$$

with $x = (F, A, \tau_1) \in \mathbb{R}^3$, $y \in \mathbb{R}$, $E_s \in \mathbb{R}$, $0 < a < 1$ is given by (5) and m is a positive integer. Note that in (21), K_m is the solution of (7) thanks to the weak sensibility of the solution with respect to this variable (see Section 5.1 for the sensibility study of the force versus K_m). We introduce the change of variables ϕ :

$$\begin{cases} \phi : \mathbb{R}^3 \rightarrow \mathbb{R}^3 \\ x \mapsto \phi(x) = [h(x), L_f(h(x)), L_f^2(h(x))]. \end{cases} \quad (22)$$

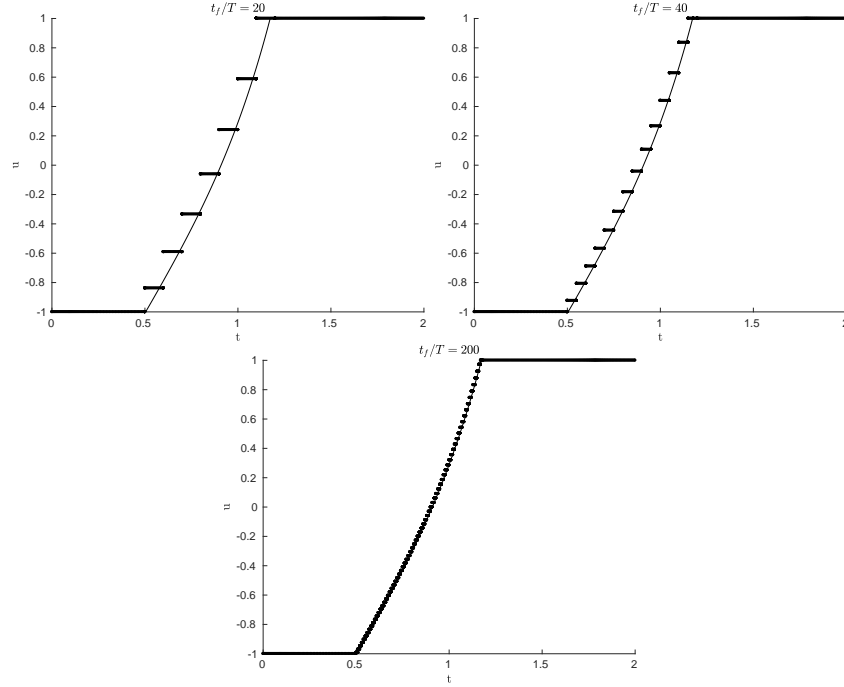


FIGURE 1. Time evolution of the permanent control (thin continuous line) and sampled-data control for several values of the sampling period $T_s \in \{T/20, T/40, T/200\}$.

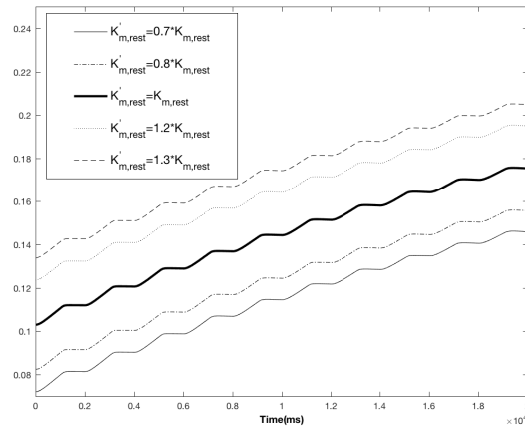


FIGURE 2. Evolution of K_m for different initial conditions (case of $I = 10ms$).

We have:

$$\frac{\partial \phi}{\partial x} = \begin{pmatrix} 1 & 0 & 0 \\ -1/(\tau_1 + \tau_2 a) & a & F/(\tau_1 + \tau_2 a)^2 \\ \frac{\partial L_f^2 h(x)}{\partial x_1} & \frac{\partial L_f^2 h(x)}{\partial x_2} & \frac{\partial L_f^2 h(x)}{\partial x_3} \end{pmatrix}$$

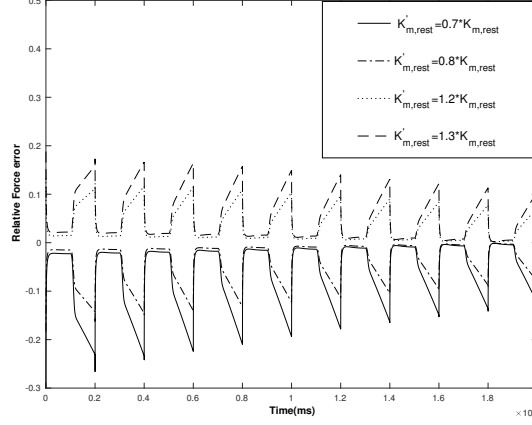


FIGURE 3. Relative error of the force for a well known and erroneous K_m initial condition (case of $I = 10ms$).

where

$$\begin{aligned}\frac{\partial L_f^2 h(x)}{\partial x_1} &= a\alpha_A + \frac{1}{(x_3 + \tau_2 a)^2} \left(1 - \frac{x_3 - \tau_{1,rest}}{\tau_{fat}} + 2\alpha_{\tau_1} x_1\right), \\ \frac{\partial L_f^2 h(x)}{\partial x_2} &= -a \frac{\tau_{fat} + x_3 + \tau_2 a}{\tau_{fat}(x_3 + \tau_2 a)}, \\ \frac{\partial L_f^2 h(x)}{\partial x_3} &= \frac{1}{(x_3 + \tau_2 a)^2} \left(x_2 a - \frac{x_1}{\tau_{fat}}\right) - \frac{2x_1}{(x_3 + \tau_2 a)^3} \left(1 - \frac{x_3 - \tau_{1,rest}}{\tau_{fat}} + \alpha_{\tau_1} x_1\right)\end{aligned}$$

and

$$\det\left(\frac{\partial \phi}{\partial x}\right) = \frac{a(x_2 a \tau_{fat}(x_3 + \tau_2 a) - \tau_{fat} x_1 + 2x_1(x_3 - \tau_{1,rest} - \alpha_{\tau_1} \tau_{fat} x_1))}{\tau_{fat}(x_3 + \tau_2 a)^3} \quad (23)$$

A sensibility study in (6) and (8) concerning respectively $-(A - A_{rest})/\tau_{fat}$ and $-(\tau_1 - \tau_{1,rest})/\tau_{fat}$, shows that neglecting these two terms induces a maximum error of 7% for A and τ_1 , and 6% for the force value. Hence we use the simplified model:

$$\frac{dA}{dt} = \alpha_A F \quad (24)$$

$$\frac{d\tau_1}{dt} = \alpha_{\tau_1} F. \quad (25)$$

Hence (4) gives:

$$\det\left(\frac{\partial \phi}{\partial x}\right) = \frac{a(x_2 a(x_3 + \tau_2 a) - x_1 - 2\alpha_{\tau_1} x_1^2)}{(x_3 + \tau_2 a)^3}. \quad (26)$$

Since $E_s(t) \geq 0$, we have:

$$\forall t \in [0, T] : (C_N, x_1) \geq 0, (x_2, x_3, K_m) > 0 \Rightarrow (x_3 + \tau_2 a)^3 > 0.$$

The equation $\det\left(\frac{\partial\phi}{\partial x}\right) = 0$ yields

$$a = 0 \Rightarrow C_N = 0 \quad (C_N = 0 \Rightarrow x_1 = 0) \quad (\text{period of rest}) \quad (27)$$

$$x_2 a(x_3 + \tau_2 a) - x_1 - 2\alpha_{\tau_1} x_1^2 = 0 \Rightarrow x_1 = \frac{1 \mp \sqrt{1 + 8\alpha_{\tau_1} x_2 a(x_3 + \tau_2 a)}}{4\alpha_{\tau_1}}. \quad (28)$$

Since $\alpha_{\tau_1} \sim 10^{-5}$, $0 \leq a \leq 1$ and $(x_3 + \tau_2 a) \sim 10^2$, we have from (28)

$$8\alpha_{\tau_1} x_2 a(x_3 + \tau_2 a) \ll 1 \Rightarrow \left(x_1 = 0 \quad (C_N = 0), x_1 = \frac{1}{2\alpha_{\tau_1}} \right).$$

Thus $(\frac{\partial\phi}{\partial x}(\hat{x}(t)) = 0, \hat{x} = (\hat{F}, \hat{A}, \hat{\tau}_1)$ for:

- $(C_N, x_1) = 0$ which corresponds to the period of rest,
- $x_1 = 1/(2\alpha_{\tau_1})$, $x_1 \sim 10^5$ which is greater than the maximum force value.

In the simplified model (4),(24) and (25), $\frac{\partial\phi}{\partial x}(\hat{x}(t))$ is invertible during stimulation period ($C_N, x_1 > 0$), and the matrix is well-conditioned.

Based on the simplified model, the modified high-gain observer is defined as:

$$\dot{\hat{x}}(t) = a^m f_1(\hat{x}(t), E_s(t)) - a^m \left(\frac{\partial\phi}{\partial x}(\hat{x}(t))^{-1} S_\theta^{-1} C^T (C\hat{x}(t) - y(t)) \right). \quad (29)$$

The experimental choice of m depends on the pulses frequency. S_θ is a symmetric definite positive matrix given by the following Lyapunov equation:

$$\theta S_\theta(t) + A^T S_\theta(t) + S_\theta(t) A = C^T C \quad (30)$$

where θ is a high-gain tuning parameter introduced in [11], and

$$A = \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{pmatrix}, \quad C = (1 \quad 0 \quad 0).$$

Hence the components of $S_\theta = [S_\theta(l, k)]_{1 \leq l, k \leq 3}$ have the following form:

$$S_\theta(l, k) = (-1)^{l+k} \binom{l+k-2}{k-1} \theta^{-(l+k-1)}, \quad \binom{n}{k} = \frac{n!}{(n-k)!k!}. \quad (31)$$

5.3. High-gain observer convergence proof. Let

$$\begin{aligned} \dot{x}(t) &= a^m(t) f_1(x(t), E_s(t)) = f(x(t), E_s(t)) \\ y(t) &= h(x(t)) \end{aligned} \quad (32)$$

and

$$\begin{aligned} z_1 &= h(x), \\ \dot{z}_1 &= L_f h(x) = a^m L_{f_1} h(x) = a^m z_2, \\ \dot{z}_2 &= L_f(L_{f_1} h(x)) = a^m L_{f_1}(L_{f_1} h(x)) = a^m z_3, \\ &\vdots \\ \dot{z}_{n-1} &= a^m z_n, \\ \dot{z}_n &= a^m L_{f_1}^n h(x) = \varphi_n(u, z). \end{aligned}$$

Then

$$\begin{pmatrix} \dot{z}_1 \\ \dot{z}_2 \\ \vdots \\ \dot{z}_n \end{pmatrix} = a^m \begin{pmatrix} 0 & 1 & \cdots & 0 \\ 0 & \ddots & \ddots & \vdots \\ \vdots & \ddots & \ddots & 1 \\ 0 & \cdots & \cdots & 0 \end{pmatrix} \begin{pmatrix} z_1 \\ z_2 \\ \vdots \\ z_n \end{pmatrix} + \begin{pmatrix} 0 \\ \vdots \\ 0 \\ \varphi_n(u, z) \end{pmatrix}. \quad (33)$$

Therefore

$$\begin{cases} \dot{z} = a^m Az + \varphi(u, z) \\ y = Cz. \end{cases} \quad (34)$$

Take $C = (1 \ 0 \ \cdots \ 0)$ and consider the system (33). We make the following assumptions.

5.3.1. Assumptions.

- H1) $\varphi_n(u, z)$ is globally Lipschitz with respect to z and uniformly with respect to u ,
- H2) Let U be the control domain, a compact $K \subset \mathbb{R}^n$ and two positive constants a_{\min} and a_{\max} such that: for all u valued in U and initial conditions $z(0) \in K$, we have $a_{\min} \leq a(t) \leq a_{\max}$,
- H3) $a(\cdot)$ is C^1 (see Remark 2 below),
- H4) $m \dot{a}(t)/a(t) < 1$.

The observer has the following expression:

$$\dot{\hat{z}}(t) = a(t)^m A \hat{z}(t) - a(t)^m S_\theta^{-1} C^T (C \hat{z}(t) - y(t)) \quad (35)$$

where S_θ is the solution of the Lyapunov equation:

$$\theta S_\theta + A^T S_\theta + S_\theta A - C^T C = 0. \quad (36)$$

The solution of (36) can be rewritten as

$$S_\theta = \frac{1}{\theta} D_\theta S_1 D_\theta. \quad (37)$$

and

$$D_\theta = \text{diag} \left(1, \frac{1}{\theta}, \cdots, \frac{1}{\theta^n} \right) \quad (38)$$

with S_1 is the solution of (36) for $\theta = 1$. Let $\rho = a^m$ and $e = \hat{z} - z \Rightarrow \dot{e} = \dot{\hat{z}} - \dot{z}$, then

$$\begin{aligned} \dot{e} &= \rho A \hat{z} + \hat{\varphi}(u, \hat{z}) - S_\theta^{-1} C^T (C \hat{z}(t) - y(t)) - \rho A z - \varphi(u, z) \\ &= \rho (A - S_\theta^{-1} C^T C) e + \hat{\varphi} - \varphi \end{aligned} \quad (39)$$

Write $\bar{e} = \rho D_\theta e$, then

$$\begin{aligned} \dot{\bar{e}} &= \dot{\rho} D_\theta e + \rho D_\theta \dot{e} \\ &= \dot{\rho} D_\theta e + \rho D_\theta (\rho (A - S_\theta^{-1} C^T C) e + \hat{\varphi} - \varphi) \\ &= \dot{\rho} D_\theta e + \rho D_\theta (\rho (A - \theta D_\theta^{-1} S_1^{-1} D_\theta^{-1} D_\theta C^T C D_\theta) e + \hat{\varphi} - \varphi), \quad D_\theta C^T C D_\theta = C^T C \\ &= \frac{\dot{\rho}}{\rho} \bar{e} + \theta \rho A \bar{e} - \theta S_1^{-1} C^T C \bar{e} + \rho D_\theta (\hat{\varphi} - \varphi), \quad D_\theta A D_\theta^{-1} = \theta A \end{aligned} \quad (40)$$

Consider the Lyapunov function: $V = \bar{e}^T S_1 \bar{e}$. Then

$$\begin{aligned}\dot{V} &= 2\bar{e}^T S_1 \dot{\bar{e}} \\ &= 2\bar{e}^T S_1 \left(\frac{\dot{\rho}}{\rho} \bar{e} + \rho \theta A \bar{e} - \theta S_1^{-1} C^T C \bar{e} + \rho D_\theta (\hat{\varphi} - \varphi) \right)\end{aligned}\quad (41)$$

$2\bar{e}^T S_1 A \bar{e} = \bar{e}^T (S_1 A + A^T S_1) \bar{e}$ and $S_1 A + A^T S_1 = -S_1 + C^T C$.

Then, we have

$$\begin{aligned}\dot{V} &= -\theta \rho V + \theta \rho \bar{e}^T C^T C \bar{e} - 2\bar{e}^T C^T C \bar{e} + 2\bar{e}^T S_1 \frac{\dot{\rho}}{\rho} \bar{e} + 2\bar{e}^T S_1 \rho D_\theta (\hat{\varphi} - \varphi) \\ &= -\theta \rho V + (\theta \rho - 2) \|C \bar{e}\|^2 + 2\frac{\dot{\rho}}{\rho} V + 2\bar{e}^T S_1 \rho D_\theta (\hat{\varphi} - \varphi) \\ &= -(\theta \rho - 2\frac{\dot{\rho}}{\rho}) V + (\theta \rho - 2) \|C \bar{e}\|^2 + 2\bar{e}^T S_1 \rho D_\theta (\hat{\varphi} - \varphi)\end{aligned}\quad (42)$$

We deduce that

$$\begin{aligned}\theta \rho - 2\frac{\dot{\rho}}{\rho} > 0 &\Rightarrow \theta > 2\frac{\dot{\rho}}{\rho^2}, \\ \theta \rho - 2 < 0 &\Rightarrow \theta < \frac{2}{\rho} \Rightarrow 2\frac{\dot{\rho}}{\rho^2} < \theta < \frac{2}{\rho} \\ &\Rightarrow \left(\frac{\dot{\rho}}{\rho} = m \frac{\dot{a}}{a} \right) < 1.\end{aligned}\quad (43)$$

Using (43) and assumption H1:

$$\begin{aligned}\dot{V} &\leq -\gamma V + 2\bar{e}^T S_1 \rho D_\theta \|\hat{\varphi} - \varphi\|, \quad \gamma > 0 \\ &\leq -\gamma V + 2\bar{e}^T S_1 \frac{\rho}{\theta^n} |\hat{\varphi}_n - \varphi_n| \\ &\leq -\gamma V + 2\bar{e}^T S_1 \frac{\rho}{\theta^n} K_1 \bar{e} \\ &\leq -\left(\gamma - \frac{2a^m}{\theta^n} K_1\right) V.\end{aligned}\quad (44)$$

For m and θ sufficiently large: $\gamma > 2a^m/\theta^n K_1 \Rightarrow \dot{V} \leq -\gamma_1 V, \quad \gamma_1 > 0$.

Remark 2. In the force-fatigue model, $a(t)$ is piecewise smooth, the lack of regularity is numerically bypassed by the choice of the integer m . For example, for $I = 10ms$, $m = 3$ is sufficient to estimate the whole variables. However, for $I = 25ms$, m must be at least equal to 5 (see observer simulations in Section 7).

6. Model predictive control (MPC). MPC computes a sequence of predicted controls to optimize plant behavior (see Fig. 4).

Model Predictive Heuristic Control (MPHC) was introduced by Richalet et al. [17] using impulse response type dynamic model. Dynamic Matrix Control (DMC) followed in 1980 (Cutler et al. [5]) using step response type dynamic model. State space formulation of MPC was introduced by Li et al. [16].

Consider a system written as

$$\dot{x} = f(x, u), \quad x(0) = x_0 \quad (45)$$

and a general cost function to be minimized

$$J_\infty(u(\cdot), x_0) = \int_0^\infty q(x^u(\tau, x_0), u(\tau)) d\tau \quad (46)$$

with

$$x^u(t, x_0) = x_0 + \int_0^t f(x^u(\tau, x_0), u(\tau)) d\tau \quad (47)$$

and

$$\begin{aligned} q(0, 0) &= 0, \quad q(\cdot) \in C^2 \\ q(x, u) &\geq c_q(\|x\|^2 + \|u\|^2), \quad c_q > 0 \\ u &\mapsto q(x, u) \text{ is convex for all } x. \end{aligned} \quad (48)$$

Without constraints, Bellman's principle of optimality (1950) gives the solution. In the case of constrained problem, instead of J_∞ , we minimize the following cost:

$$J(u(\cdot), x_0, T) = \int_0^T q(x^u(\tau, x_0), u(\tau)) d\tau + v(x^u(T, x_0)) \quad (49)$$

with

$$J^*(x_0, T) = \min_{u(\cdot)} J(u(\cdot), x_0, T) \quad (50)$$

and

$$u^*(t, x_0, T) = \arg \min_{u(\cdot)} J(u(\cdot), x_0, T). \quad (51)$$

This optimization over a finite horizon follows the algorithm:

1. Solve $\min J(u(\cdot), x_0, T)$ and find $u^*(\cdot, x_0, T)$.
2. Apply $u^*(\cdot, x_0, T)$ for $\tau \in [0, T_s]$, $0 \leq T_s \leq T$, (T_s : sampling period).
3. Repeat using $x(T_s)$ instead of x_0 .

6.1. Discrete linear system (basic case). To explain the method, we consider the discrete linear system:

$$\begin{cases} x(k+1) = Ax(k) + Bu(k) \\ y(k) = Cx(k) \end{cases} \quad (52)$$

where $x(t) \in \mathbb{R}^n$, $u(t) \in \mathbb{R}^m$, $y(t) \in \mathbb{R}^p$ are respectively the state, input and output vectors, and $t = kT_s$. This system (66) results from a discrete system modelling or continuous-discrete transformation. Practical considerations of implementation make this form more appropriate in the framework of MPC.

We suppose that the system is both controllable and observable, and we denote by $x(k+i/k)$, $i \geq 0$, the predicted state vector $x(k+i)$ with initial condition $x(kT_s)$.

Consider the unconstrained problem and the quadratic cost:

$$\begin{aligned} J(k) &= \sum_{j=k}^{\infty} \left[(y(j+1/k) - y_{ref})^T Q (y(j+1/k) - y_{ref}) \right. \\ &\quad \left. + (u(j/k) - u_{ref})^T R (u(j/k) - u_{ref}) + \Delta u(j/k)^T S \Delta u(j/k) \right] \end{aligned} \quad (53)$$

with $Q = Q^T \succ 0$, $R = R^T \succeq 0$ and $S = S^T \succeq 0$. The couple (y_{ref}, u_{ref}) is solution of (52) and $\Delta u(k) = u(k) - u(k-1)$. The solution of this problem is obtained using the LQ controller.

In the case of constrained problem, LQR could not be used to solve control problem. System (52) allows to calculate $x(k+i/k)$, $i = 1, \dots, N_r$ at time kT_s , (N_r being the receding horizon length):

$$x(k+i/k) = A^i x(k/k) + \sum_{j=0}^{i-1} A^{i-j-1} B u(k+j/k). \quad (54)$$

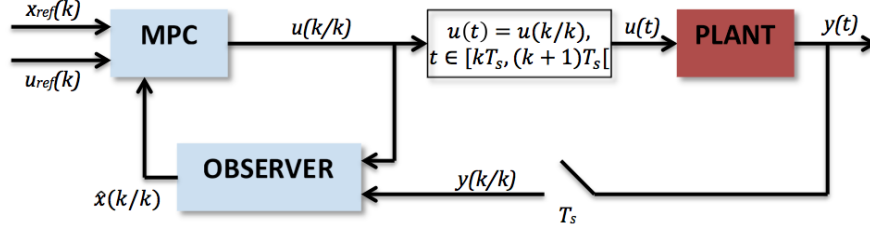


FIGURE 4. General MPC strategy diagram.

The input-output relation along the receding horizon is then:

$$\begin{pmatrix} y(k+1/k) \\ y(k+2/k) \\ \vdots \\ y(k+N_r/k) \end{pmatrix} = \begin{pmatrix} CA \\ CA^2 \\ \vdots \\ CA^{N_r} \end{pmatrix} x(k/k) + \begin{pmatrix} CB & 0 & \cdots & 0 \\ CAB & CB & \ddots & \vdots \\ \vdots & \ddots & \ddots & 0 \\ CA^{N_r-1}B & \cdots & CAB & CB \end{pmatrix} \begin{pmatrix} u(k/k) \\ u(k+1/k) \\ \vdots \\ u(k+N_r-1/k) \end{pmatrix}. \quad (55)$$

Write (55) as

$$\bar{y} = \Psi x(k/k) + \Gamma \bar{u} \quad (56)$$

where $\bar{y}_k \in \mathbb{R}^{pN_r}$, $\bar{u}_k \in \mathbb{R}^{mN_r}$, $\Psi \in \mathbb{R}^{pN_r \times n}$, $\Gamma \in \mathbb{R}^{pN_r \times mN_r}$.

Constraints can take the form:

- Control variable:

$$u_{\min} \leq u(k) \leq u_{\max}. \quad (57)$$

- Change rate of the control variables:

$$\Delta u_{\min} \leq \Delta u(k) \leq \Delta u_{\max} \quad (58)$$

with

$$\Delta u(k) = u(k) - u(k-1). \quad (59)$$

- Soft output variables constraints (relaxed output constraints using large slack variable s_v to avoid constraints conflicts when solving control problem):

$$y_{\min} - s_{v1} \leq y(k) \leq y_{\max} + s_{v1}. \quad (60)$$

- Soft state variables constraints (for the same reason as output constraints):

$$x_{\min} - s_{v2} \leq x(k) \leq x_{\max} + s_{v2} \quad (61)$$

The cost function (53) becomes:

$$\begin{aligned} J(k) = & \sum_{j=k}^{k+N_r-1} \left[(y(j+1/k) - y_{ref})^T Q (y(j+1/k) - y_{ref}) \right. \\ & \left. + (u(j/k) - u_{ref})^T R (u(j/k) - u_{ref}) + \Delta u(j/k)^T S \Delta u(j/k) \right] \end{aligned} \quad (62)$$

subject to (56) and a set of constraints among (57) – (61).

This minimization problem takes the form:

$$\begin{aligned} & \min 1/2 X^T E X + X^T F \\ & \text{subject to} \quad \bar{y} = \Psi x(k/k) + \Gamma \bar{u} \\ & \quad \quad \quad M X \leq \gamma \end{aligned} \quad (63)$$

and $\bar{y} = C\bar{x}$ with $X = [\bar{x}, \bar{u}]^T$. Optimization problem (63) is a convex Quadratic Programming (QP) problem. Denote $X^* = [\bar{x}^*, \bar{u}^*]^T$ the global minimizer at each iteration. Once \bar{u}^* is calculated, only $u^*(k/k)$ is applied at time $t = kT_s$ and we iterate the algorithm.

In the case of NMPC (non-linear Model Predictive Controller), the dynamics is defined using a discretization of the equation (45). Using (62), optimization problem becomes:

$$\begin{aligned} & \min 1/2 X^T E X + X^T F \\ & \text{subject to} \quad x(k+i/k) = g(x(k+j/k), u(k+j/k)), j = 0, \dots, N_r - 1 \\ & \quad \quad \quad M X \leq \gamma \end{aligned} \quad (64)$$

with $X = [\bar{x}, \bar{u}]^T$. Equality constraint in the optimization problem (64) is non-linear. Now convexity condition in (64) is not guaranteed, then X^* could be a local minimizer. NMPC (non-linear Model Predictive controller) is used to solve this problem.

Numerical solutions are computed using Active Set, Primal-Dual or SQP methods (Wang [19], Fletcher [10] and Boyd and Vandenberghe [4]).

In the case of force-fatigue model, one must consider the more general case of a varying T_s ($T_s \rightarrow I(i)$), then $u(i) = [I(i) \eta(i)]^T$, where $I(i) = t(i) - t(i-1)$ is the interpulse between two successive pulses and $\eta(i)$ is the pulse amplitude applied at time t_i . In this case, $\bar{u}(k) = [u(k/k) \dots u((k+N_r-1)/k)]$ is the discrete receding horizon control vector (of length N_r) to be calculated by the NMPC at time t corresponding to iteration k . Then:

$$\bar{u}(k) = \begin{pmatrix} I(k/k) & I(k+1/k) & \dots & I((t(N_r))/k) \\ \eta(t_k/k) & \eta((t_k + I(k/k))/k) & \dots & \eta((t_k + \sum_{j=0}^{N_r-1} I(k+j))/k) \end{pmatrix} \quad (65)$$

$t(N_r)$ being the final time of the optimization horizon which is unknown a priori. For instance, using single move strategy, we have $u(k/k) = u(k+1/k) = \dots = u((k+N_r-1)/k)$, (see Fig. 5).

The force-fatigue model being non-linear, an NMPC is used to solve the problem with Non-linear Programming Algorithm (NPA). In this case, criterion to be minimized is:

$$J(k) = \sum_{j=k}^{k+N_r-1} (F(j+1/k) - F_{ref})^2 \quad (66)$$

subject to: $0 \leq \eta(i) \leq 1$ and $0.01\text{ms} \leq I(i) \leq 0.1\text{ms}$.

The estimation of the state variables vector (by the high-gain observer) is used as an initial variables vector to perform the NMPC over the horizon N_r .

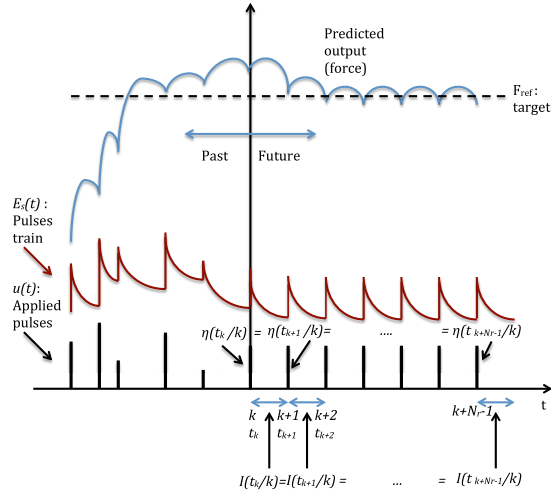


FIGURE 5. (*left half plane*) E_s and force profile for applied amplitude and interpulse stimulation, (*right half plane*) Predicted E_s and force using single move strategy to be optimized.

ALGORITHM

1. Give Final_t , $k = 1$
2. Compute $C_N(t_k)$, $F(t_k)$, $\hat{A}(t_k)$, $\hat{\tau}_1(t_k)$, $K_m(t_k)$
3. $\text{Data}_F = N_r \times \frac{u_k(1)}{\text{step}_{int}}$
with: $u_i(1) = u_k(1)$ for $i = k, k+1, \dots, k+(N_r-1)$,
and step_{int} is a submultiple of $u_i(1) = I_i$.
4. $F_{mean} = 1/\text{Data}_F \sum_{j=1}^{\text{Data}_F} F(t_k + j \text{Step}_{int}, E_s)$
with:

$$E_s = \frac{1}{\tau_c} \sum_{i=1}^{k+(N_r-1)} \left(u_i(2) H(t_{N_r} - t_i) R_i \exp\left(-\frac{t_{N_r} - t_i}{\tau_c}\right) \right)$$

($u_i(1) = u_k(1)$, $u_i(2) = u_k(2)$ for $i = k, k+1, \dots, k+(N_r-1)$,
and $t_i = t_{i-1} + u_{i-1}(1)$, $i \geq 2$, $t_{N_r} = \sum_{i=1}^{k+(N_r-1)} I_i$).

5. $u^*(k/k) = (I_k^*, \eta_k^*) = \arg \min J(I_k, \eta_k)$, $I_k \in [I_{min}, I_{max}]$, $\eta_k \in [\eta_{min}, \eta_{max}]$.
6. if $t_{k+1} \geq \text{Final}_t \Rightarrow \text{stop}$, else, $k = k+1$, back to 2.

7. Numerical results.

7.1. High-gain observer. To exhibit the interest of the use of the power m in the non-linear observer, consider a MPC based on a non-linear observer with only amplitude as control variable. We consider also the worst case of +30% of error of K_m .

The following simulation results follow the protocol used in practice (a set of periods with stimulation and rest time slots in each period). During the stimulation time slot, the control is calculated to bring the force to F_{ref} and in the rest time slot, the stimulation amplitude is set to 0. In this section, only two stimulation periods are considered.

For $I = 10ms$ (100Hz) and $I = 25ms$ (40Hz), we take $m = 3$ and $m = 5$ respectively. Fig.6 and Fig.7 represent the \hat{A} for $I = 10ms$ and $I = 25ms$, respectively. Fig.8 and Fig.9 are the $\hat{\tau}_1$ for $I = 10ms$ and $I = 25ms$. \hat{A} converges after 50ms when $I = 10ms$ and 100ms when $I = 25ms$. Concerning $\hat{\tau}_1$, it converges after 75ms when $I = 10ms$ and 200ms when $I = 25ms$. Large I seems to delay the convergence of the observer.

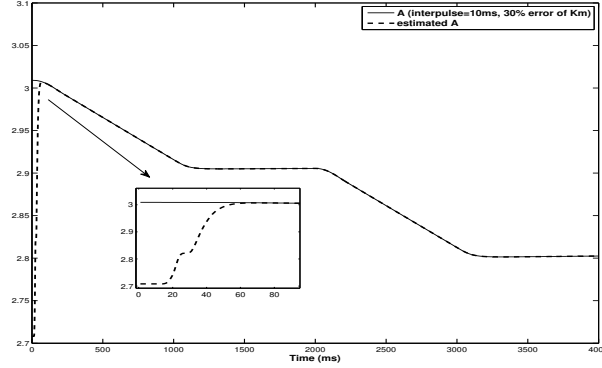


FIGURE 6. Evolution of A and \hat{A} for $I = 10$, 30% error of K_m .

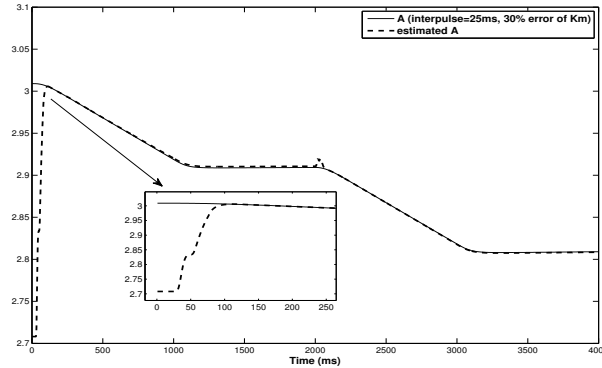
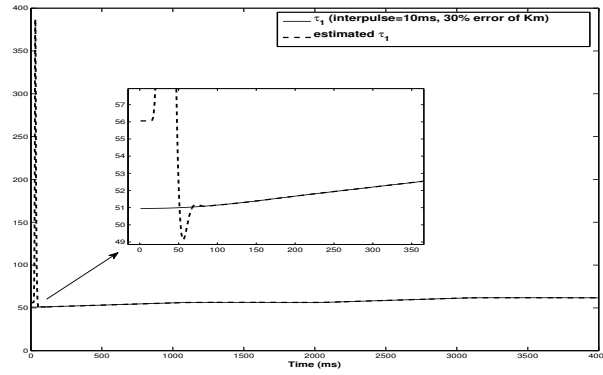
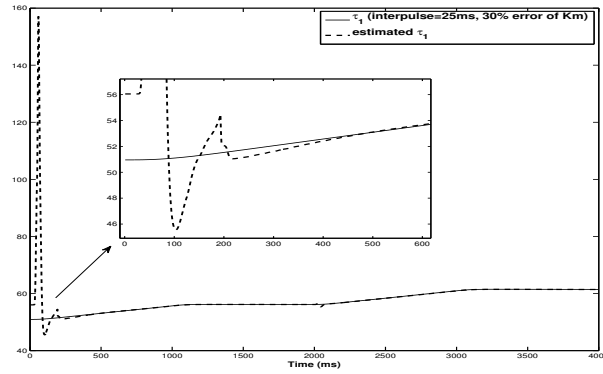
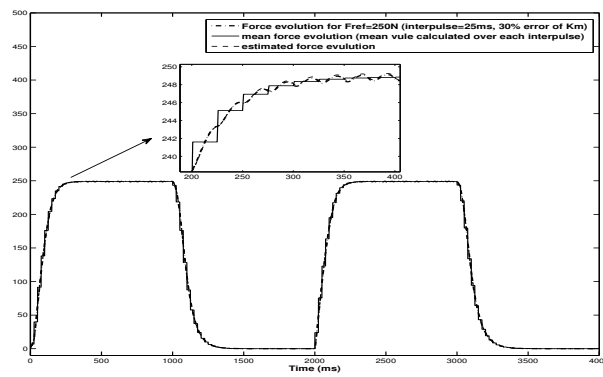


FIGURE 7. Evolution of A and \hat{A} for $I = 25$, 30% error of K_m .

Fig.10 represents the force response for amplitude control strategy (for a receding horizon $N_r = 10$) based on the proposed observer for $I = 25ms$ and a force reference of $250N$. Force mean value converges to the force reference after $200ms$.

7.2. MPC with interpulse and amplitude as control variables. In this section, five stimulations periods are presented (due to the experimental protocol). Figures 11, 12 and 13 are the force response in the case of $F_{ref} = 425N$ and ($N_r = 3, 5, 10$), the interpulse and the amplitude controls for $N_r = 10$, respectively. As expected, $N_r = 10$ gives the best regulation performances (response time and overshoot).

Before $t = 6000ms$ with $N_r = 10$, the force is correctly maintained at $F_{ref} = 425N$. For $t \geq 6000ms$ (starting from the forth period), the fatigue level A is very

FIGURE 8. Evolution of τ_1 and $\hat{\tau}_1$ for $I = 10$, 30% error of K_m .FIGURE 9. Evolution of τ_1 and $\hat{\tau}_1$ for $I = 25$, 30% error of K_m .FIGURE 10. Evolution of F , \hat{F} and F mean value over I for $I = 25$, 30% error of K_m , $F_{ref} = 250N$.

high so that the maximum value of the amplitude control (see Fig. 13) and the interpulse control (see Fig. 12) cannot maintain F at F_{ref} . Maximum interpulse frequency is not used at the beginning of the fourth and the fifth periods. Higher

value of N_r could correct this problem (supposing that global minimizer of MPC algorithm is reached at each iteration). However, increasing N_r will render MPC algorithm time consuming and cause a problem for real time implementation.

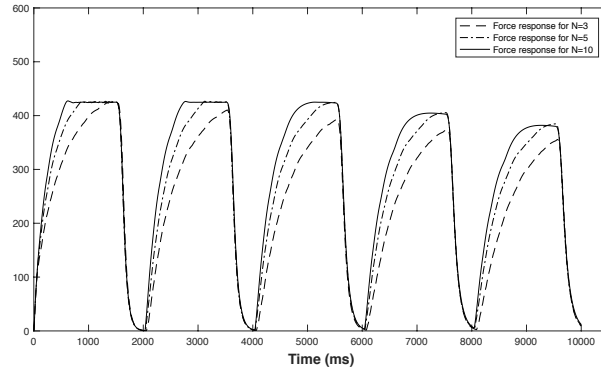


FIGURE 11. Evolution of the force for a reference force of $425N$ and different receding horizons (3, 5 and 10)

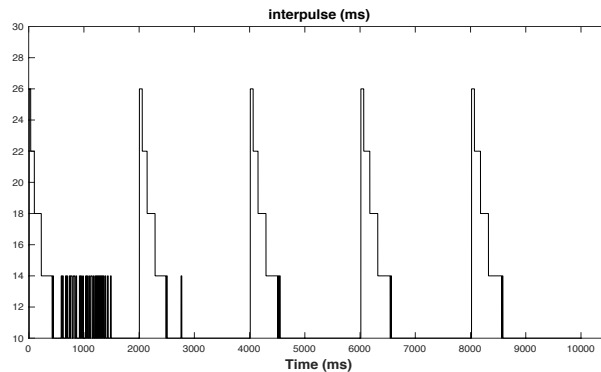


FIGURE 12. Evolution of the interpulse (control) for a reference force of $425N$ and a predictive horizon of 10.

8. Conclusion. This work deals with the control and estimation of the state variables of the Ding et al. force-fatigue model where the control is the interpulse and/or amplitude of the electrical stimulation. Preliminary geometric analysis of the force control controlling force level is presented with the aim of a future PMP control strategy. In the case of a fixed interpulse, the proposed high-gain observer using the force measurements exhibits the relation between the interpulse and the parameter m to perform accurate variables estimation. Model Predictive Control (MPC) strategy is presented in the case of both stimulation interpulse and amplitude as control variables. Simulation results are presented concerning the high-gain observer and the MPC force control. These simulations show the effect of the receding horizon on the control efficiency. Reasonable value of N_r is however suitable to guarantee a short computation time for real time implementation.

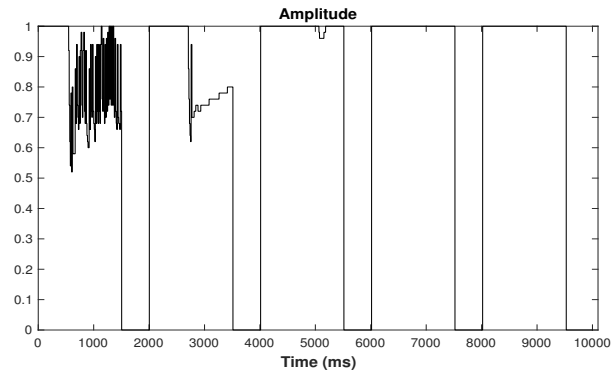


FIGURE 13. Evolution of the amplitude (control) for a reference force of $425N$ and a predictive horizon of 10.

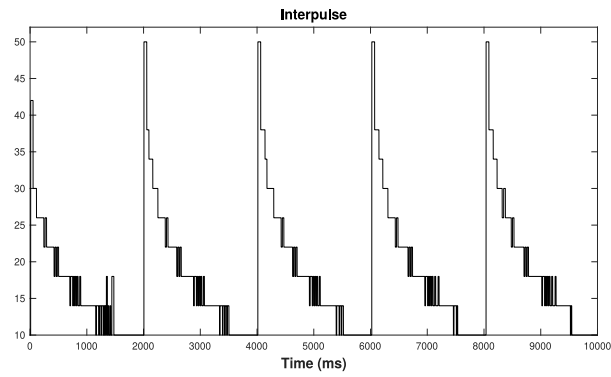


FIGURE 14. Evolution of the interpulse (control) for a reference force of $425N$ and a predictive horizon of 10.

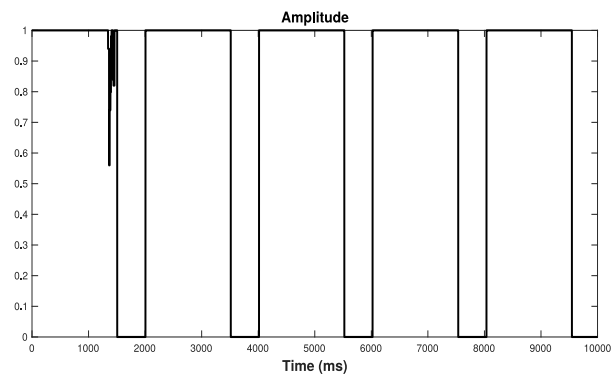


FIGURE 15. Evolution of the amplitude (control) for a reference force of $425N$ and a predictive horizon of 3.

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Received for publication April 2018.

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